



MEMS FOR RF/MICROWAVE WIRELESS APPLICATIONS: THE NEXT WAVE – PART II

Microelectromechanical systems (MEMS) technology is on the verge of revolutionizing RF and microwave applications.¹ The requirements of present day and future RF and microwave systems for lower weight, volume, power consumption and cost with increased functionality, frequency of operation and component integration are driving the development of new RF/microwave MEMS components and system architectures. Part I, published previously, presented a brief discussion of RF and microwave system requirements, and then introduced the enabling potential of MEMS to meet these requirements. Part II describes the revolutionary possibilities afforded by MEMS in system integration and novel architectures.

The interest in MEMS technology for RF and wireless applications stems from the view that, because of its flexibility, it can be exploited to overcome the limitations exhibited by integrated RF devices¹ and, in doing so, enables circuits with new levels of performance not achievable otherwise. Thus, the ultimate goal in applying RF MEMS is to propagate the device-level benefits all the way up to the system level to attain unprecedented levels of system performance, as shown in **Figure 1**. In this section,

early examples of circuits exploiting MEMS are presented.

MEMS Inductor-based Circuits

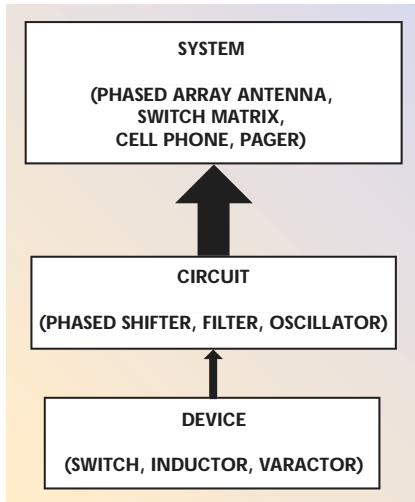
Inductors are key elements determining the performance of tuned circuits, in particular, impedance matching networks, low noise amplifiers and voltage-controlled oscillators (VCO).² Improving the gain, power dissipation or phase noise of these circuits has, in turn, led to incorporating MEMS-based on-chip inductors. Perhaps the quintessential example of the

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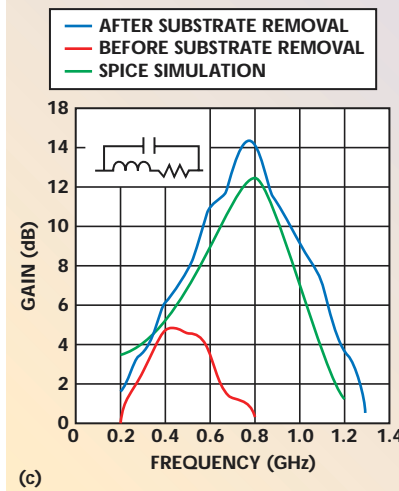
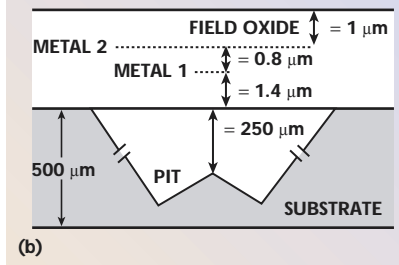
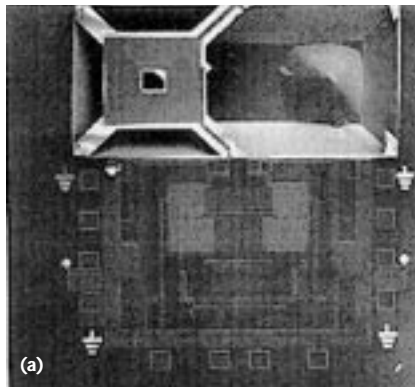
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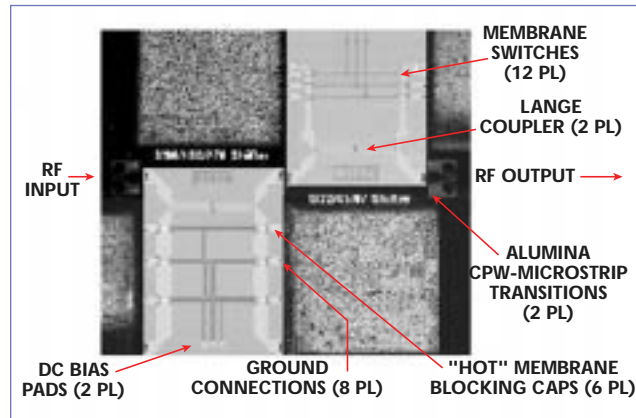
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▲ Fig. 1 Typical RF MEMS "food chain."

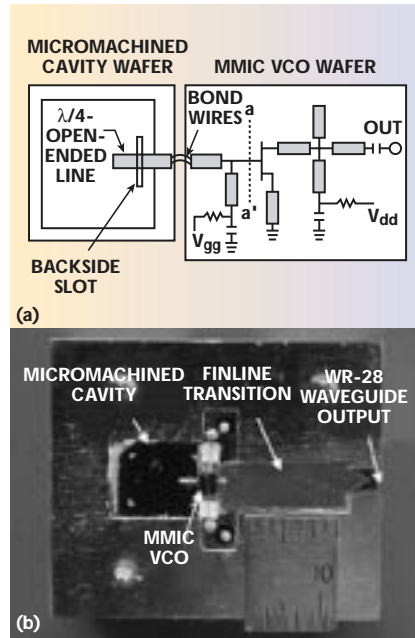


▲ Fig. 2 CMOS RF amplifier with suspended MEMS inductor (a), the inductor's cross section (b) and its performance (c).



▲ Fig. 3 X-band four-bit phase shifter.

Fig. 4 VCO with micromachined resonator cavity; (a) schematic and (b) top view of cavity oscillator assembled in a WR-28 waveguide package. ▼



improvements brought about by a MEMS inductor is the pioneering demonstration of a tuned CMOS RF amplifier by Chang, Abidi and Gaitan.³ Figure 2 shows a performance comparison obtained from amplifiers, constructed with and without the inductor's substrate removed. Improvements of approximately 12 dB in gain, and a factor of two higher center frequency capability with the MEMS inductor are evident.

MEMS Varactor-based Circuits

Varactors are essential components wherever circuit tunability is required (for example, in variable matching circuits and VCOs). However, as the move toward more compact form factors persists, and since good semicon-

ductor varactors are incompatible with conventional IC processes, a number of efforts to demonstrate the performance level enabled by MEMS varactors have been undertaken.^{4,5,6} A recent example is that shown by Dec and Suyama⁶ who, employing 1.4 pF parallel plate capacitors with a Q of 14 at 2 GHz

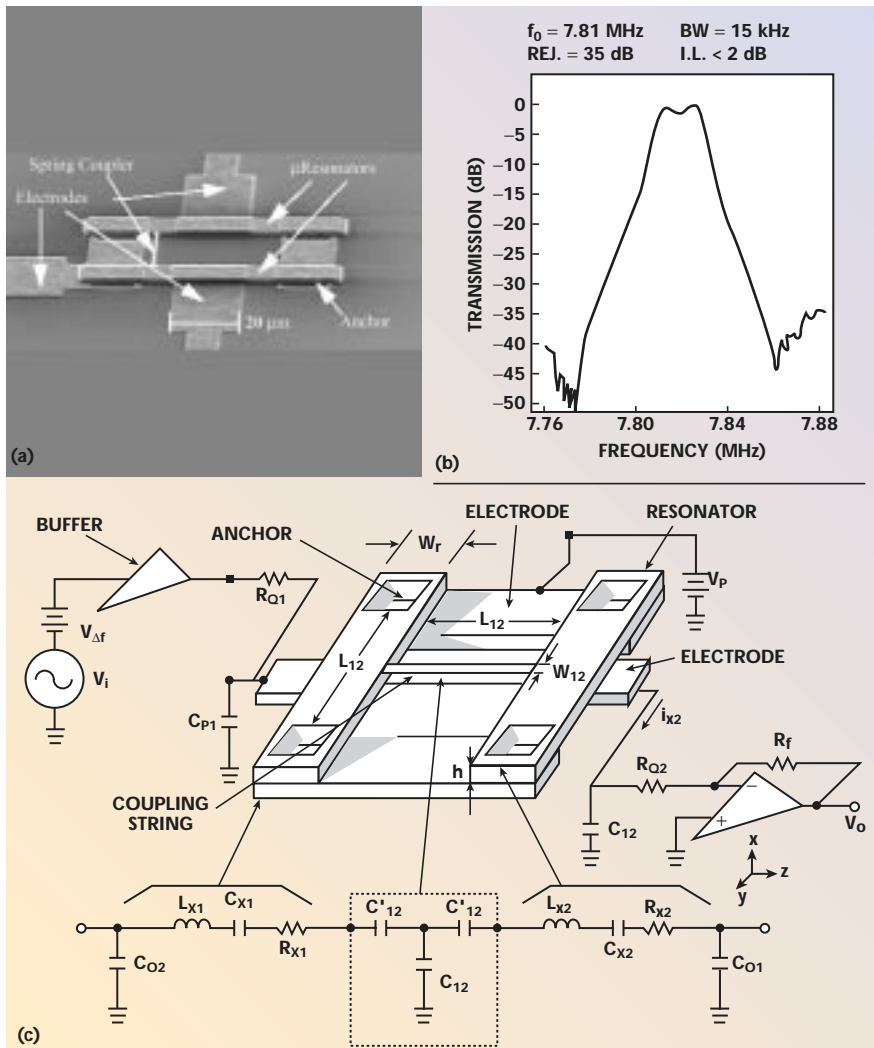
and fabricated in standard polysilicon surface micromachining technology, demonstrated a 2.4 GHz VCO with a phase noise of -122 dBc/Hz at 1 MHz offset from the carrier and a 3.4 percent frequency tunability range over a 5 V tuning voltage.

MEM Switch-based Circuits

The excellent performance of prototype MEMS switches — for instance insertion loss and isolation of approximately 0.1 dB and 50 dB, respectively, from DC to 4 GHz⁷ — has demonstrated their great potential for replacing lossy and power hungry semiconductor switches in numerous applications, including T/R switches, phaseshifters, switchable filters, cross-bar/multiplexing, tunable antennas and phased arrays. In this regard, an X-band 4-bit phaseshifter with an average insertion loss of only 1.4 dB and a return loss greater than 11 dB, has been recently demonstrated,⁸ as shown in Figure 3.

Micromachined Cavity Resonator-based Circuits

It is well-known that the Q of a cavity resonator is proportional to its volume. Therefore, it is natural to consider the utilization of cavity resonators in circuits and applications, such as emerging millimeter-wave commercial applications, whose frequency and performance levels cannot be met otherwise. These circuits include oscillators, VCOs and filters. A recent example⁹ is that of a 33.2 GHz MMIC oscillator stabilized by a micromachined cavity, shown in Figure 4, which exhibited a phase noise of -113 dBc/Hz at 1 MHz offset, an 18 dB improvement over its MMIC free-running counterpart.



▲ Fig. 5 Two-resonator MEM filter (a), performance (b) and its circuit model (c).

Micromechanical Resonator-based Circuits

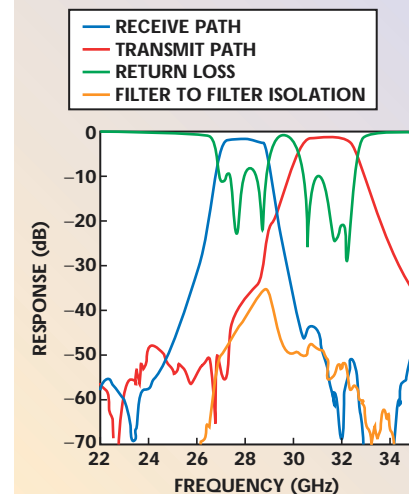
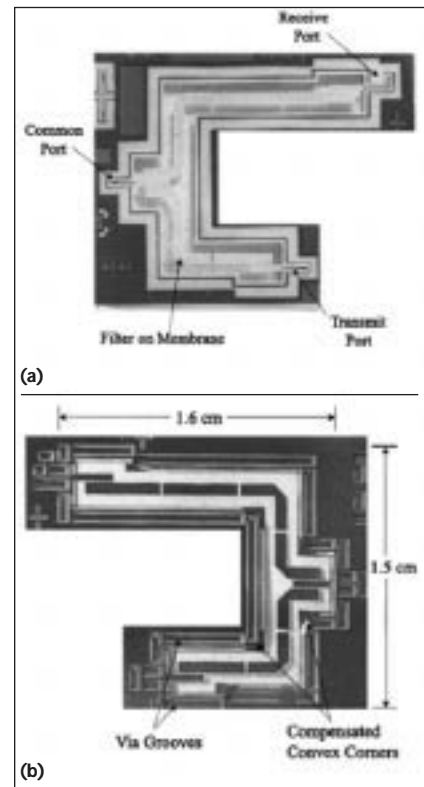
At lower frequencies, cavity resonators become impractical due to their excessively large dimensions. Micromechanical resonators, in turn, become very attractive because their resonance frequency is proportional to the square root of their stiffness-to-mass ratio.¹ Thus, considerable effort, most notably by Nguyen,¹⁰ continues to be spent on developing MEM resonators using filters as demonstration vehicles. Currently, the resonance frequency capabilities are well below 1 GHz (at 156 MHz) and Qs measured under vacuum conditions approach 9400.¹¹

The design of MEM-resonator-based filters essentially proceeds in the electrical domain along conventional lines, except that use is made of electro-mechanical analogies to cross over from the electrical proto-

type model to the mechanically realizable structure.¹ Thus far, filters operating at frequencies as low as a few kilohertz to several megahertz have been demonstrated. **Figure 5** shows the equivalent circuit^{12,13} and measured response of a two-resonator filter operating at 7.8 MHz.

Micromachined Transmission Line-based Circuits

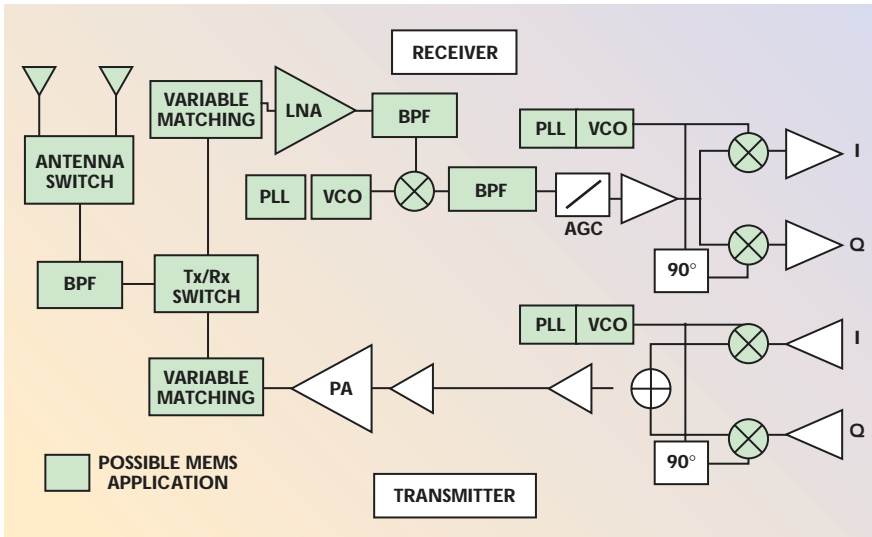
Transmission lines are rather versatile components in RF and microwave electronics, as they are key elements of many circuits and systems, in addition to embodying circuits in themselves. A number of circuits that employ various types of micromachined lines have been demonstrated (namely, filters and diplexers,¹⁴ and antennas).¹⁵ Photographs and performance of some of these circuits are shown in **Figure 6**.



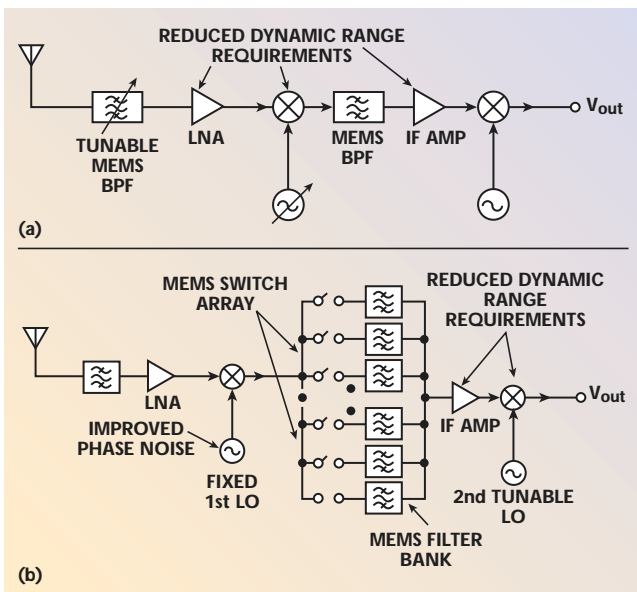
▲ Fig. 6 Fabricated micromachined K-band diplexer's (a) top and (b) bottom views and (c) frequency response, and (d) micromachined waveguide-based G-band (140 to 220 GHz) antenna.

RF/MICROWAVE MEMS SYSTEM INTEGRATION AND NOVEL ARCHITECTURES

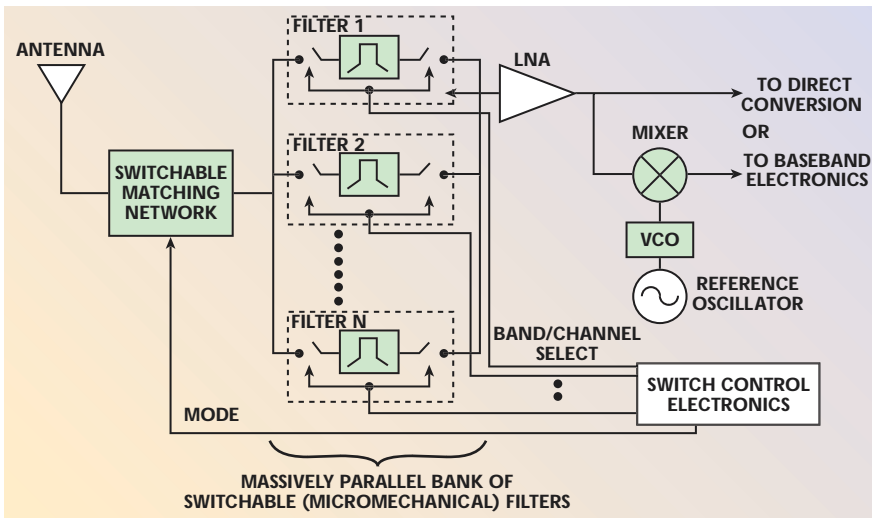
With the increasing trend toward more powerful portable wireless systems, the need for integration is now more acute than ever. Indeed, in applications such as cellular, personal communications systems (PCS), wireless networking, radar, LMDS, MMDS, steerable antennas and satellite communications, the need for reduced parts count, form factor and power consumption, and lighter weight, remains paramount. One of the promises of MEMS is that, through the superior levels of RF and microwave component performance it enables, great strides can be made in the areas of system integration and novel architectures, directed at fulfilling these needs.



▲ Fig. 7 Simplified conventional transceiver architecture.



▲ Fig. 8 MEMS-based receiver (a) employing a front-end MEMS tunable filter and (b) with acoustic resonant IF filter banks.



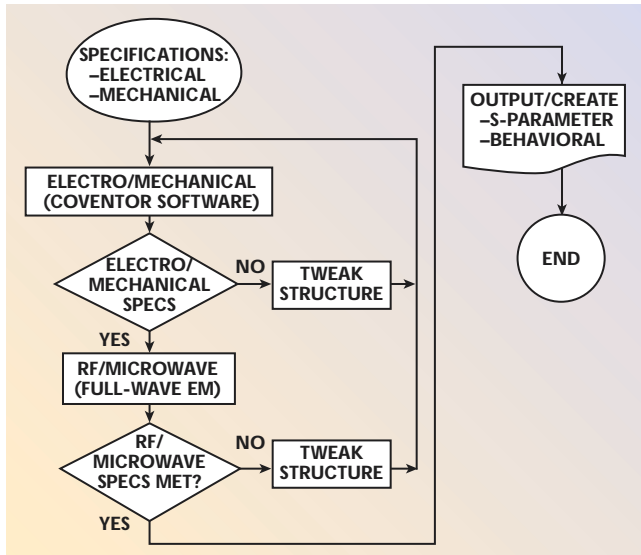
▲ Fig. 9 Switchable receiver front-end architecture.

Approaches for RF MEMS Insertion in Wireless Systems

Two approaches to RF MEMS insertion are possible — bottoms-up and top-down considerations. In the bottoms-up approach, one would proceed by employing straightforward or direct component replacement as dictated by an established system architecture. In the top-down scenario, one would begin by devising a system architecture, not prejudiced by the usual limitations imposed by conventional RF components, that would exploit to the highest degree possible their RF MEMS realizations.

A typical example of the bottoms-up approach is shown in **Figure 7** by a conventional transceiver architecture. In this context, it is clear that the off-chip passives components, switches, filters, VCOs, mixers, oscillators and diplexers are all candidates for direct replacement by their MEMS counterparts. Another example of the bottoms-up approach would be that of a phased array antenna (PAA) system. Indeed, since MEM switches require nano-joule-type switching energy, and virtually zero standby power, it is no longer a fantasy to envision PAAs containing orders of magnitude more elements than currently feasible, by the direct replacement of the conventional switches.

On the other hand, a number of proposals, predicated upon the top-down approach, have already appeared in literature,^{16,17} with particular emphasis on transceivers. For example, Larson has proposed two receiver concepts shown in **Figure 8**.¹⁶ One employs a MEMS-based tunable front-end filter aimed at improving integrability through the drastic lowering of the IF enabled by the filter. The other receiver employs an acoustic resonant IF filter bank aimed at simplifying system realization by eliminating the need for a tunable first LO, that is, by replacing it with a fixed LO and exploiting the switchable filter bank



▲ Fig. 10 RF MEMS device design flow.

to affect IF band selection. Along the same lines, Nguyen proposed a front-end architecture,¹⁷ in which band selection is accomplished by the orchestrated action of a switched matching network and a massively parallel bank of MEMS filters shown in **Figure 9**.

Enabling The Next Wave: Concurrent System Design

Without an appropriate environment, the design and timely implementation and time-to-market deployment of MEMS-based systems would be virtually hopeless, due to the design dimensions added by the multi-physics/multi-domain nature of MEMS devices. Thus, MEMS design tools are essential to the successful commercialization of MEMS-based RF/wireless systems.

CAD tools¹⁸ provide an environment where virtually all aspects of MEMS device design and production, from first principles physical simulation to system-level device models to packaging to manufacturing process feasibility, can be concurrently addressed to avoid last-minute surprises.

A typical design flow for an RF MEMS device is shown in **Figure 10**. From mechanical and electrical specifications, the device is designed iteratively¹⁹ until both sets of specifications are satisfied. Within the iteration loops, the software may be used to explore, and correct for, the impact of a number of potential influences on device performance/yield, such as package stress and strain state, manufacturing process or temperature variability, and hermeticity. The design process ends with an S-parameter file and/or a behavioral reduced order model for subsequent utilization in a circuit/system simulator.

CONCLUSION

In this two-part article, a succinct tutorial on the field of MEMS with particular emphasis on their ability to enable the breakthroughs that will launch the next class of RF/microwave wireless systems applications was presented. In Part I, MEMS-based RF devices, such as inductors, varactors, switches, resonators and transmission lines, their fabrication, packaging and a CAD design methodology, were progres-

sively introduced. In Part II, some of their most important circuit and system applications, and possible approaches for their insertion, were described. In conclusion, the need for a concurrent system design environment, such as that provided by currently available software, was suggested, and a typical RF MEMS device design flow that could exploit such an environment was presented. ■

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Figure Attributions

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Figures 2a/b: J.Y.-C. Chang, A.A. Abidi and M. Gaitan, "Large Suspended Inductors on Silicon and Their Use in a 2 μm CMOS RF Amplifier," *IEEE Electron Device Letters*, Vol. 14, 1993, pp. 246–248, Figure 2.

Figure 2c: J.Y.-C. Chang, A.A. Abidi and M. Gaitan, "Large Suspended Inductors on Silicon and Their Use in a 2 μm CMOS RF Amplifier," *IEEE Electron Device Letters*, Vol. 14, 1993, pp. 246–248, Figure 4.

Figure 3: A. Malczewski, S. Eshelman, B. Pil-lans, J. Ehmke and C.L. Goldsmith, "X-band RF MEMS Phase Shifters for Phased Array Applications," *IEEE Microwave Guided Wave Letters*, Vol. 9, No. 12, December 1999, pp. 517–519, Figure 2.

Figure 4a: Y. Kwon, C. Cheon, N. Kim, C. Kim, I. Song and C. Song, "A Ka-band MMIC Oscillator Stabilized with a Micromachined Cavity," *IEEE Microwave and Guided Wave Letters*, Vol. 9, No. 9, September 1999, pp. 360–362, Figure 1.

Figure 4b: Y. Kwon, C. Cheon, N. Kim, C. Kim, I. Song and C. Song, "A Ka-band MMIC Oscillator Stabilized with a Micromachined Cavity," *IEEE Microwave and Guided Wave Letters*, Vol. 9, No. 9, September 1999, pp. 360–362, Figure 2.

Figure 5a: F.D. Bannon III, J.R. Clark and C.T.-C. Nguyen, "High Frequency Micro-electromechanical IF Filters," *Technical Digest*, IEEE International Electron Devices Meeting, San Francisco, CA, December 8–11, 1996, pp. 773–776, Figure 1.

Figure 5b: F.D. Bannon III, J.R. Clark and C.T.-C. Nguyen, "High Frequency Micro-electromechanical IF Filters," *Technical Digest*, IEEE International Electron Devices Meeting, San Francisco, CA, December 8–11, 1996, pp. 773–776, Figure 2.

Figure 5c: C.T.-C. Nguyen, "Microelectro-mechanical Devices for Wireless Communica-tions," Proceedings, 1998 IEEE International Microelectromechanical Systems Workshop, Heidelberg, Germany, January 25–29, 1998, pp. 1–7, Figure 7.

Figures 6a/b: A.R. Brown and G.M. Rebeiz, "A High-performance Integrated-band Diplexer," *IEEE Transactions on Microwave Theory and Techniques*, Vol. 47, No. 8, August 1999, pp. 1477–1481, Figure 6.

Figure 6c: A.R. Brown and G.M. Rebeiz, "A High-performance Integrated-band Diplexer," *IEEE Transactions on Microwave Theory and Techniques*, Vol. 47, No. 8, August 1999, pp. 1477–1481, Figure 10.

Figure 6d: J.W. Digby, C.E. Collins, B.M. Towlson, et al., "Integrated Micro-machined Antenna for 200 GHz Operation," *IEEE MTT-S Digest*, Figure 2.

Figure 8a: L.E. Larson, "Microwave MEMS Technology for Next-generation Wireless Communications – Invited Paper," *1999 IEEE MTT-S Digest*, Figure 4.

Figure 8b: L.E. Larson, "Microwave MEMS Technology for Next-generation Wireless Communications – Invited Paper," *1999 IEEE MTT-S Digest*, Figure 5.

Figure 9: C.T.-C. Nguyen, "Micromechanical Components for Miniaturized Low Power Communications," *Pro. 1999 IEEE MTT-S International Microwave Symposium RF MEMS Workshop (on Microelectromechanical Devices for RF Systems: Their Construction, Reliability and Applications)*, Anaheim, CA, June 18, 1999, pp. 48–77.